



Astro 500



Techniques of Modern Observational Astrophysics

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Lecture Outline

Part II. Detectors *continued*

- S/N formulation
- S/N regimes
- DQE
- Photon propagation method

- NIR detectors
- MIR & FIR detectors

Signal

- Point source

- We are measuring photon flux

- $E(\gamma) = f(\gamma) A t$

A is
telescope
collecting
area; t is
exposure
time

- Resolved source

- We are measuring surface brightness

- $E(\gamma) = I(\gamma) A \Omega t$

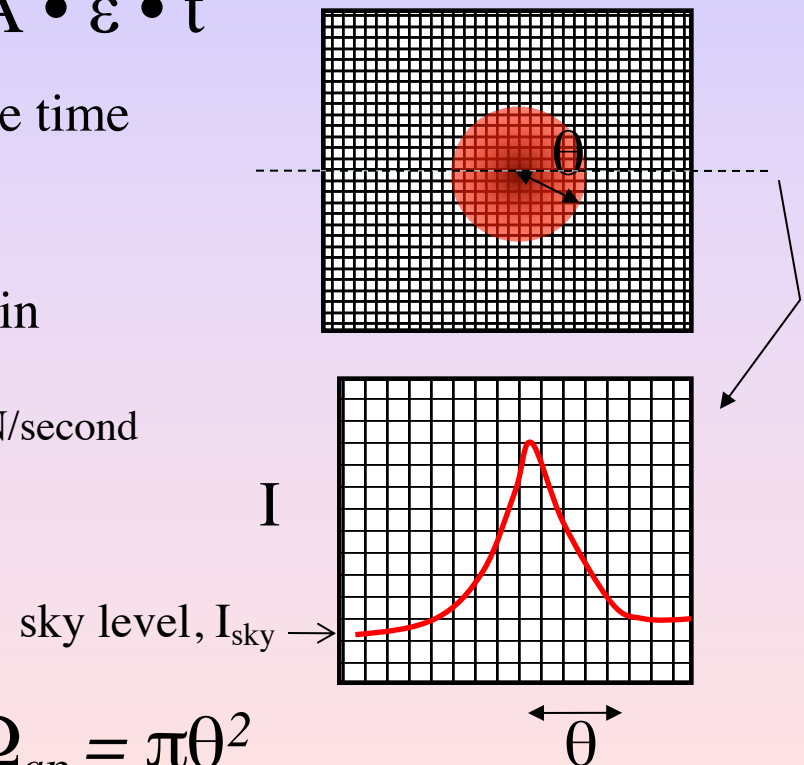
$S(\gamma)$

This ignores any inefficiencies in the measurement process

Signal-to-Noise (S/N)

- $$\text{Signal} = \underbrace{S_{obj}}_{\substack{\text{detected e-/second: } S_{obj} = S_{DET} \cdot \text{gain} \\ \text{DN/second}}} \cdot \underbrace{\varepsilon \cdot t}_{\substack{\text{total system efficiency} \\ \text{exposure time}}} = f_{obj} \cdot A \cdot \varepsilon \cdot t$$

- Consider the case where we count all the detected e- in a circular aperture with radius θ . In this case $\Omega_{ap} = \pi\theta^2$



Aside: how big an area do we want to integrate over?

Noise Sources

$\pi\theta^2$ must be in units that match I_{sky} (e.g., flux per unit solid angle)

$\pi\theta^2$ must be in pixel units

$$\sqrt{S_{obj} \cdot \varepsilon \cdot t} = \sqrt{f_{obj} \cdot A \cdot \varepsilon \cdot t} \Rightarrow \text{shot noise from source}$$

$$\sqrt{I_{sky} \cdot A \cdot \pi\theta^2 \cdot \varepsilon \cdot t} \Rightarrow \text{shot noise from sky in aperture of circular radius } r$$

$$\sqrt{RN^2 \cdot \pi\theta^2} \Rightarrow \text{readout noise in aperture of circular radius } r$$

$$\sqrt{[RN^2 + (0.5 \times \text{gain})^2] \cdot \pi\theta^2} \Rightarrow \text{more general RN}$$

$$\sqrt{\text{Dark} \cdot \pi\theta^2 \cdot t} \Rightarrow \text{shot noise in dark current in aperture of circular radius } r$$

In general:
replace $\pi\theta^2$ with Ω_{ap} where Ω_{ap} is the solid angle of your measurement aperture.

$$S_{obj} \cdot \varepsilon = e^- / \text{sec} \text{ from the source}$$

If I_{sky} is in units of photon flux per pixel, then $I_{sky} \cdot \varepsilon = e^- / \text{sec} / \text{pixel}$
from the sky and $S_{sky} = I_{sky} \cdot A \cdot \pi\theta^2 = I_{sky} \cdot n_{pix}$

If I_{sky} is in units of photon flux per unit solid angle $d\Omega$ (e.g., arcsec^{-2}), then $S_{sky} = I_{sky} \cdot \pi\theta^2$ but express $\pi\theta^2$ in the same units of solid angle.

RN = read noise (as if $RN^2 e^-$ had been detected)

$\text{Dark} = e^- / \text{sec} / \text{pixel}$

NB Ω_{ap} must have suitable units (pixels for RN and Dark and typically arcsec^2 for I_{sky}).

S/N for object measured in aperture with radius θ :

$$n_{\text{pix}} = \# \text{ of pixels in the aperture} = \pi\theta^2$$

$$\frac{\text{Signal}}{\text{Noise}} = \frac{S_{\text{obj}}}{\left[\underbrace{f_{\text{obj}} \cdot A \cdot \epsilon \cdot t}_{\text{Noise from object } e^- \text{ in aperture}} + \underbrace{I_{\text{sky}} \cdot A \cdot \Omega_{\text{ap}} \cdot \epsilon \cdot t}_{\text{Noise from sky } e^- \text{ in aperture with solid angle } \Omega_{\text{ap}}} + \underbrace{\left(RN + \frac{\text{gain}}{2} \right)^2 \cdot n_{\text{pix}} + \text{Dark} \cdot t \cdot n_{\text{pix}}}_{\text{Readnoise in aperture with angle } \Omega_{\text{ap}} \text{ solid expressed in pixels}} \right]^{\frac{1}{2}}}$$

S_{obj} and S_{sky} are the signal components.
 $f_{\text{obj}} \cdot A \cdot \epsilon \cdot t$ is the signal from the object.
 $I_{\text{sky}} \cdot A \cdot \Omega_{\text{ap}} \cdot \epsilon \cdot t$ is the signal from the sky.
 $\left(RN + \frac{\text{gain}}{2} \right)^2 \cdot n_{\text{pix}} + \text{Dark} \cdot t \cdot n_{\text{pix}}$ is the noise from the dark current in aperture with angle Ω_{ap} solid expressed in pixels.

All the noise terms added in quadrature

Note: always calculate in e^- why?

What is ignored in this S/N eqn?

Telescope
diameter



- Explicit inclusion of collecting aperture (i.e., $A \propto D_{\text{tel}}^2$)
- Break-out of terms that go into total system efficiency (starting from the top of the atmosphere)
 - Bias level/structure correction and errors
 - Flat-fielding correction and errors
 - Charge Transfer Efficiency (CTE) 0.99999/pixel transfer
 - Non-linearity when approaching full well
 - Scale changes in focal plane
 - Interpolation errors and correlation

S/N regimes

- Two basic regimes:
 1. Photon-limited (shot-noise from source+sky photons)
 2. Detector-limited (read-noise)
- In photon-limited case, two important sub-regimes
 - a. Source-limited
 - b. Sky-limited

S/N regimes: limiting cases

Let's assume CCD with Dark=0, well sampled read noise.

$$S / N = \frac{f_{obj} \cdot A \cdot \varepsilon \cdot t}{\left[f_{obj} \cdot A \cdot \varepsilon \cdot t + I_{sky} \cdot A \cdot \Omega_{ap} \cdot \varepsilon \cdot t + (RN)^2 \cdot n_{pix} \right]^{\frac{1}{2}}}$$

Note: seeing or source-size comes in with Ω_{ap} and n_{pix} terms

1a. Bright Sources: $(S_{obj} \varepsilon t)^{1/2}$ dominates noise term

$$S/N \approx \frac{S_{obj} \varepsilon t}{\sqrt{S_{obj} \varepsilon t}} = \sqrt{S_{obj} \varepsilon t} \propto t^{\frac{1}{2}}$$

S/N limiting cases (*contd*)

$$S/N = \frac{f_{obj} \cdot A \cdot \varepsilon \cdot t}{\left[f_{obj} \cdot A \cdot \varepsilon \cdot t + I_{sky} \cdot A \cdot \Omega_{ap} \cdot \varepsilon \cdot t + (RN)^2 \cdot n_{pix} \right]^{\frac{1}{2}}}$$

$I_{sky,pix} = I_{sky} \Omega_{ap} / n_{pix}$,
i.e., this is the sky
flux per pixel

1b. Sky Limited:

$$(\sqrt{I_{sky,pix} \varepsilon t} > 3^* RN)$$

$$S/N \propto \frac{S_{obj} \varepsilon t}{\sqrt{I_{sky} A \Omega_{ap} \varepsilon t}} \propto t^{\frac{1}{2}}$$

Note: seeing
comes in with
 Ω_{ap} or n_{pix} term

2. Read-noise Limited:

$$(\sqrt{I_{sky,pix} \varepsilon t} < 3^* RN)$$

$$S/N \propto \frac{S_{obj} \varepsilon t}{RN \sqrt{n_{pix}}} \propto t$$

What does this imply
about exposure time?

*Why 3? How about 1?

S/N regimes: limiting cases

Again, let's assume CCD with Dark=0, well sampled read noise.

$$S / N = \frac{f_{obj} \cdot A \cdot \varepsilon \cdot t}{\left[f_{obj} \cdot A \cdot \varepsilon \cdot t + I_{\text{sky}} \cdot A \cdot \Omega_{ap} \cdot \varepsilon \cdot t + (RN)^2 \cdot n_{pix} \right]^{\frac{1}{2}}}$$

But now let's take into account the explicit dependencies not just on time but on collecting area A and measurement aperture Ω_{ap} .

S/N limiting cases (*contd*)

1a. Bright Sources:

$$S/N \approx \frac{S_{obj} \varepsilon t}{\sqrt{S_{obj} \varepsilon t}} \propto (A \varepsilon t)^{\frac{1}{2}} \propto (\varepsilon t)^{\frac{1}{2}} D_{tel}$$

1b. Sky Limited:

$$S/N \propto \frac{S_{obj} \varepsilon t}{\sqrt{I_{sky} \Omega_{ap} \varepsilon t}} \propto (A \varepsilon t / \Omega_{ap})^{\frac{1}{2}} \propto (\varepsilon t)^{\frac{1}{2}} D_{tel} / \theta_{ap}$$

Could choose 1

2. Read-noise Limited: ($\sqrt{I_{sky} \varepsilon t} < 3 \times RN$)

$$S/N \propto \frac{S_{obj} \varepsilon t}{\sqrt{n_{pix} RN^2}} \propto \varepsilon t \left(D_{tel}^2 / \theta_{ap} \right)$$

D_{tel} : telescope diameter θ_{ap} : measurement aperture radius

DQE

- DQE is often defined as the *effective quantum efficiency* of a CCD relative to an ideal detector with no read-noise. In the source-limited regime, ignoring dark-current:

$$DQE = QE / \left[1 + \frac{RN^2}{QE \cdot S_{obj} \cdot t} \right]$$

where QE is the CCD quantum efficiency.

- This can be generalized for any noise-regime, and including dark-current.
- A related concept is the *effective system efficiency*, DQE_{sys} , of which CCD QE is only one part.

How are these quantities formulated?

S/N regimes (recap)

$$S/N = \frac{S_{obj} \cdot \varepsilon \cdot t}{\left[S_{obj} \cdot \varepsilon \cdot t + I_{sky} \cdot \Omega_{ap} \cdot \varepsilon \cdot t + (RN)^2 \cdot n_{pix} \right]^{\frac{1}{2}}}$$

1. Photon Limited:

$$S/N \propto t^{\frac{1}{2}}$$

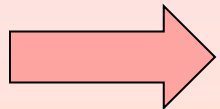
2. Read-noise Limited:

$$S/N \propto t$$

What does this imply
about exposure time?

Photon propagation

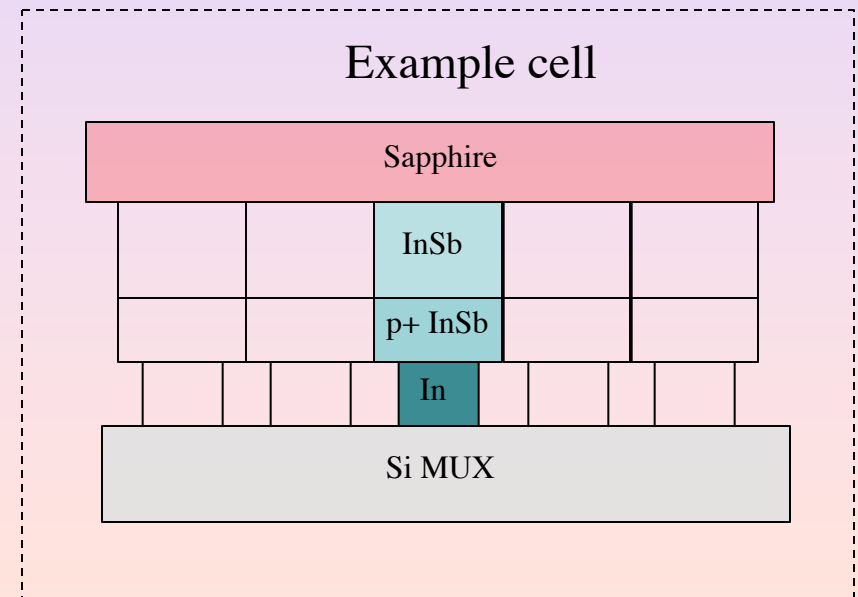
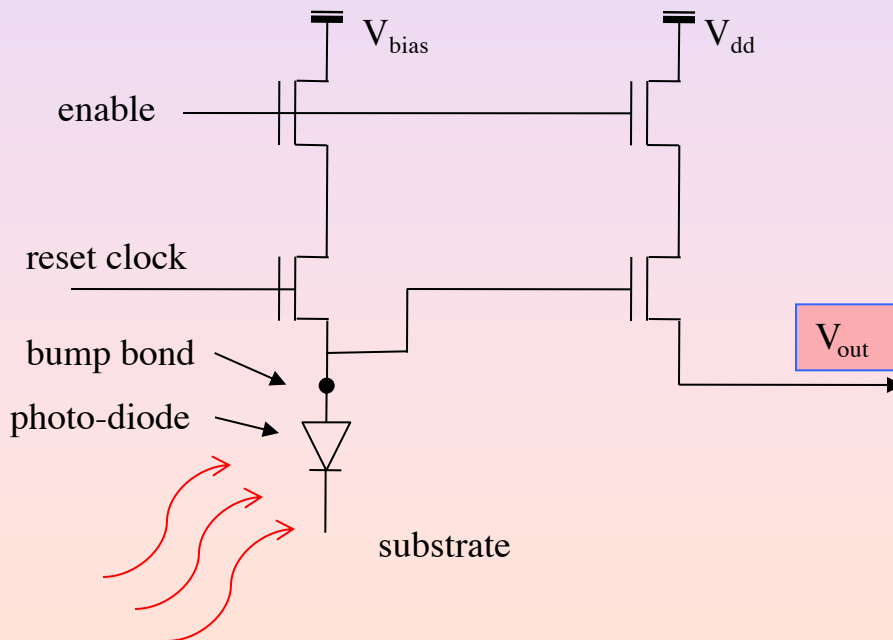
- Gain and read-noise of a detector can be determined empirically from a set of data taken
 - With constant source over a range of exposures
 - With constant integration with a range of source flux
- Noise definition:
 - $g = \text{gain} = e^-/\text{DN}$
 - $R = \text{detector noise } (e^-)$
 - $p = \text{photon noise } (e^-) = [\text{counts}(\text{DN}) \cdot g]^{1/2}$
 - $\text{Noise}^2 (e^-) = p^2 + R^2$
 - $(\text{Noise}/g)^2 (\text{DN}) = (p/g)^2 + (R/g)^2$
 $= \text{counts} (\text{DN}) \cdot g^{-1} + (R/g)^2$



A relation existing between mean counts and standard deviation (in DN) that yields the gain (e^-/DN) and read-noise (e^- rms)

Near-Infrared Detectors

- Hybride photo-diode or photo-voltaic arrays
 - InSb or HgCdTe semi-conductors, grown on polished sapphire substrate: this is what converts photons to electrons
 - Bump-bonded with In (a conductor) to Si read-out structure (MOSFET or CMOS multiplexer)



Three primary NIR detectors

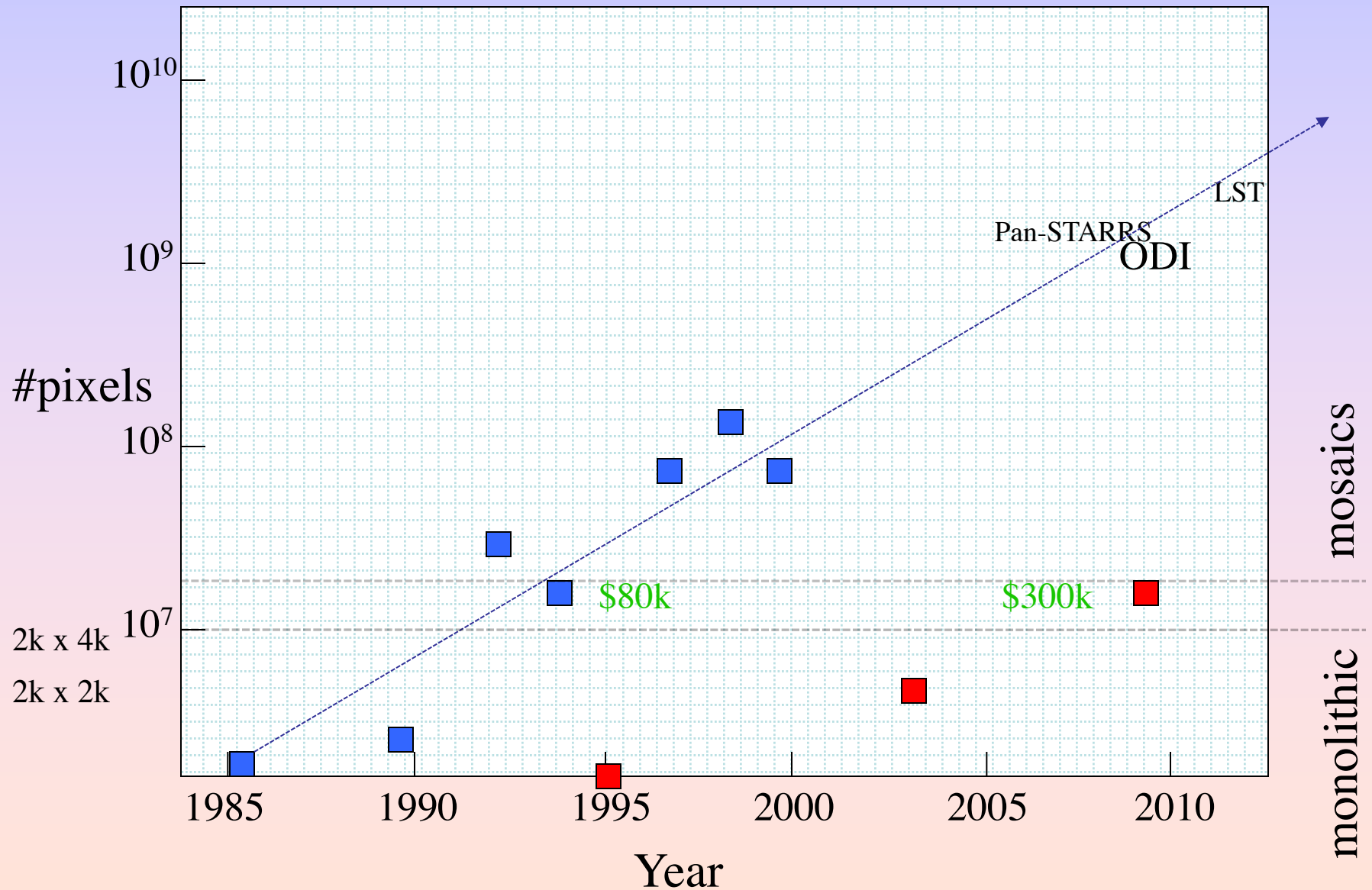
- HgCdTe: 0.5-5 micron sensitivity, QE up to 80%.
 - Blue sensitivity requires substrate thinning. Red cut-off determined by Hg:Cd mix (out to 17 micron, but sapphire only transmits to 6.5 micron). NIR spectrograph for SALT uses mix that cuts off at 1.7 microns.
- InSb: typically 0.9-5 micron, again depends on doping.
- InGaAs: 0.8-1.7 micron, possible extension to 2.2 micron. Much lower cost.
- (Also PtSi, but different physical principle and low QE – 2-3%)
 - Both semiconductors overlap with CCD wavelength sensitivity.
 - HgCdTe vendor has pushed short wavelength sensitivity.
 - InSb vendor pushed long wavelength sensitivity – into the thermal IR. Implications for instrument design: blocking, cooling, backgrounds, performance, etc.
 - Historically InSb's had 30% higher QE (80 vs 60), but this appears no longer to be true.

Near-Infrared Detectors

- **Format:**
 - 64x64 in 1988
 - 256x526 by 1990 (HST/NICMOS)
 - 1048x1048 in 1995
 - 2048x2048 early 2000's
 - vendors taking orders for 4096x4096
- **Pixel size:**
 - originally 76 micron/pix; now typically 18-27 micron/pix (CCDs have 3-24 micron/pix)
- **Dynamic range:**
 - 3×10^5 to 10^6 e- full well – comparable to CCDs.
- **QE:** comparable to CCDs
- **Dark current:** typically much higher than for CCDs
- **Read-noise:**
 - 400 e- circa 64x64
 - 40 e- by mid-90's 1024x1024 arrays
 - 15-20 e- with correlated multiple sampling

Requires more
cooling
Why?

CCD and NIR array size and \$\$\$



Differences with optical CCDs

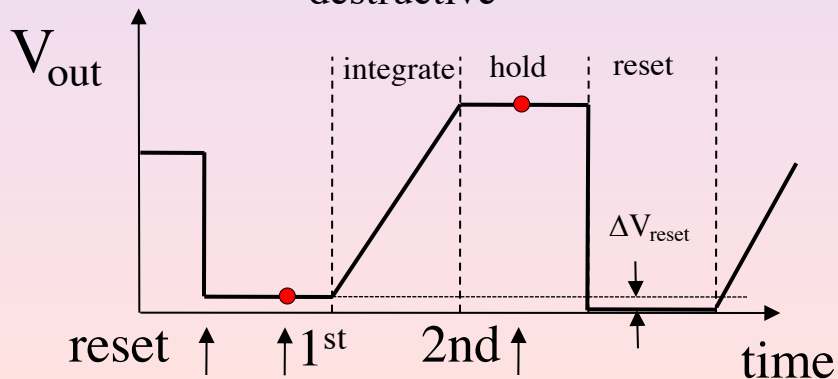
- Aside from *cost*, *size* (no longer really an issue), *read-noise* and *dark current*....
- Detectors are not charge buckets or corrals, but diodes: charge is collected, but fixed (until reset).
 - No shifting of charge along rows or columns
- Charge on each pixel is addressed via the Si multiplexer and sampled directly.
 - (Multiple) sub-arrays can be directly (and efficiently) addressed for read-out, even at different rates.
- Sampling of this charge (as a voltage) is *non-destructive*.
- Detectors are fundamentally non-linear

What does this imply for sampling, performance, and observing modes?

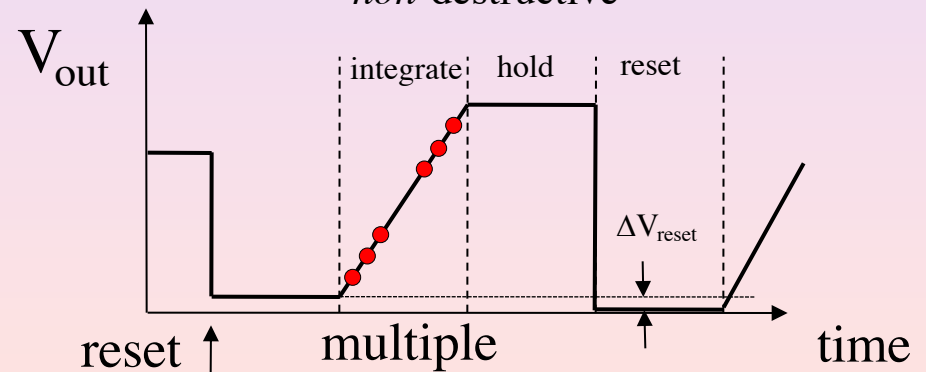
IR Array read-out modes

- As with CCDs, voltage produced by photoelectrons is sampled with an integrating circuit and digitized.
- Again, there is kTc noise associated with charge-injection when resetting the sample and integration circuitry voltages.
- Non-destructive reads allow for more than correlated double-sampling (CDS).
- Now correlated multiple sampling (CMS or *Fowler sampling*)

CDS for CCDs
destructive



CMS for CCDs
non-destructive



- Note where the voltage is sampled in CMS. *Why is this possible? Why is it advantageous? How many samples are ideal? What is the ideal temporal spacing?*

Detector non-linearity

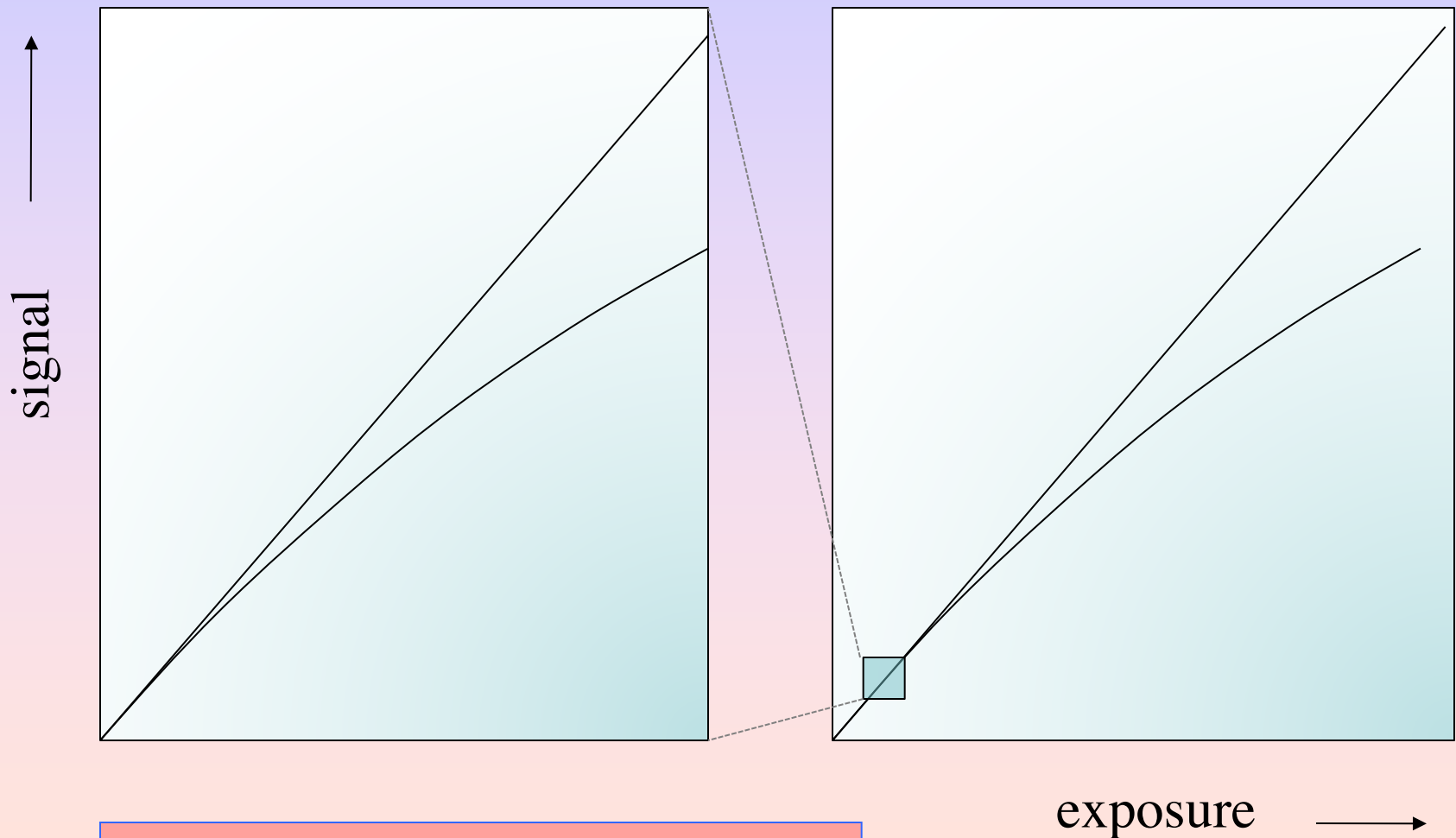
- Photo-diodes operate by having a capacitance at their pn (detector) junction.
- This capacitance depends on the (reverse) voltage across the junction.
- This voltage depends on the total number of electrons in the conduction band of the photo-diode (either thermal or photoelectrons)

 NON LINEAR

- Non-linearity must be calibrated
- Adds dimension to data gathering, reduction and calibration

Question: If NIR detectors require non-linearity corrections, how come we worry about CCD non-linearity?

Detector non-linearity



Non-linear all the way down...

Mid- and Far-Infrared Detectors

- It's all about tuning the band-gap for the right photon energy level, and then suffering the consequences of the material-sciences headaches
- (Blocked) Impurity band conductors (IBC, or BIB):
 - doped Si using Ga, As, or Sb:
 - 5-28 micron wavelength sensitivity up to 1024x1024 array size
- Extrinsic Ge semiconductors, Ga doped
 - Large photon diffusion requires large pixels (500-700 microns)
 - Examples:
 - Spitzer/MIPS (32x32 array) 70 micron sensitivity
 - Herschel/PACS - same
 - AKARI satellite: 160 micron sensitivity

Find values for WIYN & SALT instr.

- Detector gain, read-noise, system efficiency
 - WIYN
 - WHIRC
 - Bench Spectrograph
 - ODI
 - NEID
 - SALT
 - SALTCAM
 - RSS
 - HRS

Assignment:

- ❖ Work as a team
- ❖ Tabulate information
- ❖ Assess availability, ease of access
- ❖ Bring to class to present